Effects of Design Geometries and Nonlinear Losses on Gain in Silicon Waveguides with Erbium-Doped Regions

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Abstract—Silicon waveguides integrated with doped dielectric gain media may allow the design of planar light sources with electronic control. In this paper, the effects of design geometries and nonlinear losses on the gain in crystalline silicon waveguides with erbium-doped regions are investigated. We show that by using multitrench geometries, the power confinement can be increased and higher gain-to-nonlinear-loss ratio achieved. Net gain can be improved as much as 0.38 dB/cm in multitrench waveguides.

Index Terms—Erbium, integrated optics, laser amplifiers, waveguides.

I. INTRODUCTION

S ILICON light emitters are the missing links for future silicon-based photonic integration. A main challenge is to find a way to attain high quantum efficiency [1]. Many approaches have been employed to obtain optical gain in silicon, ranging from silicon nanocrystals to erbium-doped silicon, and yet there have been few achievements [2]–[5].

Conventionally, light amplification at telecommunication wavelengths has been achieved using erbium-doped low-index host materials, such as erbium-doped oxide glass (phosphosilicate and soda-lime glasses) and ceramic thin films (Al₂O₃, Y_2O_3) [6]–[8]. Well-developed fabrication processes and high erbium solubility in a small volume have attracted great interest in these host materials, and hence planar dielectric waveguides are fabricated to deliver wafer-scale light amplifiers. For instance, A1₂O₃:Er³⁺ waveguides with 0.1 dB/cm background loss have been fabricated with low-cost and reliable fabrication processes, 80 nm gain bandwidth, and up to 2.3 dB/cm amplification performance [9]–[11]. The net gain can be further enhanced to 4.1 dB/cm in similar waveguide devices using phosphate glass [12].

Silicon slot waveguides, which are submicrometer slots embedded between waveguides, have been experimentally demonstrated to provide a high optical confinement in the slot region that can accommodate possible optical gain media

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[13], [14]. Approaches that confine light in low-index structures have been presented to show achievements in CMOScompatible light-emitting devices. To achieve lasing, the gain provided by the active material in the slot region must be high enough to overcome losses in the whole waveguide. One way of achieving this goal is to use horizontal slot waveguides [15], in which slots are sandwiched between amorphous silicon (α silicon). However, the main advantage of using silicon comes from electronic integration. Thus, the use of crystalline silicon is important for enabling electronic integration.

In this paper, we analyze the structure of crystalline silicon waveguides with multiple erbium-doped Al₂O₃ regions sandwiched in between. The advantages of the proposed structures are the following. 1) Light amplification at telecommunication wavelengths is induced by erbium-doped Al₂O₃. 2) Electronic tuning capabilities such as switching and modulating are provided by the crystalline silicon platform. The theoretical study presented in this paper evaluates the gain and loss competitions in multitrench crystalline silicon waveguides with erbium-doped Al₂O₃. Especially, effects of erbium concentration, power confinement, and nonlinear losses on gain are discussed in detail. Different power distribution profiles in the waveguides and erbium concentrations up to $5 \times 10^{20} \text{cm}^{-3}$ have been considered. In particular, nonlinear losses that mitigate the optical gain, including excited-state absorption (ESA) and upconversion (UC) in erbium-doped Al₂O₃ as well as free-carrier absorption in silicon, have been analyzed. Different design geometries and the associated nonlinear losses have been discussed for their impacts on optical gain. Multitrench geometries are shown to ultimately increase the power confinement in the low-index regions and produce higher gainto-nonlinear-loss ratio. We show that net gain can be improved as much as 0.38 dB/cm in multitrench waveguides compared to single-trench waveguides.

The rest of this paper is organized as follows. First, we introduce the theoretical model used to calculate gain and losses along the device using physical parameters determined experimentally. Next, we present the device geometries used in our calculations, and, finally, we conclude with key results and discussions.

II. THEORETICAL MODEL AND ANALYSIS

To evaluate signal amplification by erbium ions, power, signal, and amplified spontaneous emissions (ASE) profiles

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along the waveguide, population inversion, and power confinement need to be analyzed. After determining signal and pump power distributions, signal evolution along the waveguide is performed by separating the signal propagation in silicon and in erbium-doped Al₂O₃. The theoretical model presented in this paper includes gain and loss mechanisms in erbiumdoped Al₂O₃ regions and silicon regions, as summarized in Fig. 1. Prior to modeling, finite element analysis was performed on each waveguide geometry to obtain normalized power distribution. Then, the analysis of pump, signal, and ASE variation in erbium-doped Al₂O₃ was carried out based on four-level rate equations, considering the energy levels presented in Fig. 2 [16]–[18]. In the silicon region, this model emphasizes free-carrier-induced nonlinear losses. After evaluating signal, pump, and ASE power in the waveguide at each propagation step, the total power is redistributed in the geometry according to the mode analysis. Given the small geometrical size and design of the waveguides used in this paper, the mode profile was found to be invariant with respect to the propagation direction. Finally, aggregate signal gains of the proposed structures are calculated from the abovementioned analysis at the end of the waveguide [19], [20]. As silicon's absorption coefficient was found to be $\geq 100 \text{ cm}^{-1}$ at shorter wavelengths, pumping at 1.48 μ m was observed to be the only feasible scheme considered in this model.

The rate equations following the energy levels presented in Fig. 2 can be expressed as [16]–[18]:

$$\frac{dN_1}{dt} = -W_{12}N_1 - R_{12}N_1 + W_{21}N_2 + R_{21}N_2 + \frac{N_2}{\tau_2} + C_2N_2^2 - C_{37}N_3^2$$
(1)

$$\frac{dN_2}{dt} = W_{12}N_1 + R_{12}N_1 - W_{21}N_2 - R_{21}N_2 - \frac{N_2}{\tau_2} + \frac{N_3}{\tau_3} - 2C_2N_2^2 - W_{24}N_2$$
(2)

$$\frac{dN_3}{dt} = -\frac{N_3}{\tau_3} + \frac{N_4}{\tau_4} - C_{37}N_3^2 \tag{3}$$

$$\frac{dN_4}{dt} = -\frac{N_4}{\tau_4} + C_2 N_2^2 + W_{24} N_{24}$$
(4)

$$N_T = N_1 + N_2 + N_3 + N_4 \tag{5}$$

$$R_{ij} = \frac{\sigma_{ij}(\nu_s)}{h\nu_s} I_s + \sum_{j=1}^m \frac{\sigma_{ij}(\nu_j)}{h\nu_j} [I_{ASE+} + I_{ASE-}]$$
(6)

$$W_{ij} = \frac{\sigma_{ij}(v_P)}{hv_P} I_P.$$
(7)

Cooperative UC and exited-state absorption were taken into account in erbium-doped Al₂O₃ by these rate equations. N_T is the total erbium concentration, N_1 , N_2 , N_3 , N_4 are the Er³⁺ concentrations in the energy levels ⁴I_{15/2}, ⁴I_{13/2}, ⁴I_{11/2}, and ⁴I_{9/2}, respectively, as shown in Fig. 2, τ_2 , τ_3 , τ_4 are the luminescence lifetimes corresponding to each energy level, respectively, and C_2 is the cooperative UC coefficient. For high erbium concentrations, second-order UC was found to play a critical role and hence was included in this model (C_{37}). However, at low erbium concentrations ($\sim 3 \times 10^{20}$ cm⁻³), if the waveguide is carefully designed and fabricated, the effect



Fig. 1. Gain and loss mechanisms in erbium-doped AL_2O_3 and silicon based on the theoretical model in the proposed structures.



Fig. 2. Energy levels for Er3+ pumped at 1.48 μ m.

of second-order UCs can be negligible and ignored ($C_{37} = 0$) [17]. R_{12} , R_{21} , W_{12} , and W_{21} are the induced signal and pump transition rates, W_{ESA} is the induced ESA rate, I_s , I_p , and $I_{ASE\pm}$ are the signal, pump, and ASE intensities, respectively, and σ_{ij} is the emission or absorption cross section. The physical parameters used for numerical analysis are from the experimental data shown in Table I. Steady-state population densities N_1 , N_2 , N_3 , and N_4 at each energy level were calculated by setting the left-hand side of (1)–(4) to zero at each propagation step. The steady-state evolution of the pump, signal, and ASE powers in erbium-doped Al₂O₃ are expressed as the following propagation equations [18]:

$$\frac{dP_p(z)}{dz} = -\gamma_p(z)P_p(z) - \alpha_p P_p(z) \tag{8}$$

$$\frac{dP_s(z)}{dz} = [\gamma_{21}(z, \nu_s) - \gamma_{12}(z, \nu_s)]P_s(z, \nu_s) - \alpha_s P_s(z)$$
(9)

$$\frac{dP_{ASE+}(z, v_j)}{dz} = [\gamma_{21}(z, v_j) - \gamma_{12}(z, v_j)]P_{ASE+}(z, v_j) + Mhv_i \Delta v_i \gamma_{21}(z, v_j) - \alpha_e P_{ASE+}(z, v_j)$$
(10)

$$+Mnv_j\Delta v_j\gamma_{21}(z,v_j) - a_s P_{ASE+}(z,v_j)$$
(10)
$$dP_{ASE-}(z,v_j)$$

$$\frac{dz_{ASE}(z, v_j)}{dz} = -[\gamma_{21}(z, v_j) - \gamma_{12}(z, v_j)]P_{ASE+}(z, v_j) - Mhv_j \Delta v_j \gamma_{21}(z, v_j) + \alpha_s P_{ASE+}(z, v_j).$$
(11)

Here, the solution is obtained for boundary values of

$$P_p(0) = P_p, P_s(0) = P_s,$$

$$P_{ASE+}(0, v_j) = P_{ASE-}(L, v_j) = 0 (j = 1, 2, ..., m)$$

where *m* is the number of frequency slots for ASE. The coefficient *M* in (10) and (11) represents the number of guided modes at the signal wavelength. The absorption and emission coefficients $\gamma_{12}(z, v_s)$, $\gamma_{21}(z, v_s)$, and $\gamma_p(z)$ are given by

$$\gamma_{12}(z,\nu_s) = \iint_A \psi_s(x,y)\sigma_{12}(\nu_s)N_1(x,y,z)dxdy \tag{12}$$

$$\gamma_{21}(z, v_s) = \iint_A \psi_s(x, y) \sigma_{21}(v_s) N_2(x, y, z) dx dy$$
(13)

$$\gamma_p(z) = \iint_A \psi_p(x, y) [\sigma_{12}(v_p) N_1(x, y, z) - \sigma_{21}(v_p) N_2(x, y, z) + \sigma_{24}(v_p) N_2(x, y, z)] dx dy$$
(14)

where $\gamma_p(x, y)$ and $\gamma_s(x, y)$ are normalized pump and signal intensity profiles in the erbium-doped regions, respectively. $\Psi_s(x, y)$ and $\Psi_p(x, y)$ are the normalized signal and pump intensity profile. *A* is the cross-sectional area of the erbiumdoped active region, and σ_{12} , σ_{21} , and σ_{24} are the absorption and emission cross sections.

With regard to free-carrier absorption in silicon, at each propagation step, $I_{Si}(x, y, z)$ represents the variation of pump intensity at each individual point $\langle x, y, z \rangle$ in silicon, and can be expressed as [21]

$$\frac{dI_{Si}(x, y, z)}{dz} = -\alpha_{fca}(z)I_{Si}(x, y, z) - \beta I_{Si}^{2}(x, y, z) \quad (15)$$

where

$$\alpha_{fca}(z) = 1.45 \times 10^{-17} \left(\frac{\lambda}{1.55}\right)^2 N(x, y, z) \ cm^{-1}.$$
 (16)

 β represents the two-photon absorption coefficient and $\alpha_{fca}(z)$ represents the free-carrier absorption expressed in [22]. Also, N(x, y, z) is the free-carrier density at each individual point in silicon and it is deduced from [21, Eq. (6)]

$$N(x, y, z) = \frac{\tau_0 \beta I_{Si}^2(x, y, z)}{2h\nu} + f(i).$$
(17)

Here, τ_0 is the free carrier recombination lifetime. The last term f(i) is the free-carrier absorption induced by current injection in the event of electronic modulation. Since electronic manipulation is not discussed in this paper, free-carrier absorption induced by carrier injection is ignored in the following numerical calculations. At each propagation step, the pump power in silicon affected by free-carrier absorption can be represented as [21]

$$P_{si}(z) = \iint_{Si} I_{si}(x, y, z) dx dy.$$
(18)

Upon determination of the pump profile along the waveguide, the free-carrier losses of the pump can be estimated as

$$\Delta P_{si} = P_{si}(z+1) - P_{si}(z).$$
(19)

Numerical calculations by pumping at 1.48 μ m are summarized as follows.

 TABLE I

 Physical Parameters Used for Numerical Analysis of Silicon

 Waveguide with Erbium-Doped Active Regions [16]–[18]

Parameter	Value
Waveguide length	1 cm
Pump wavelength	1.48 μm
Signal power	$1 \ \mu W$
Er concentration (N_{Er})	$5 \times 10^{19} - 5 \times 10^{20} \text{ cm}^{-3}$
Pump absorption cross	
section (1.48 μ m) σ_{12}	$2.7 \times 10^{-21} \text{ cm}^2$
Pump emission cross	
section (1.48 μ m) σ_{21}	$0.77 \times 10^{-21} \text{ cm}^2$
Signal absorption cross	
section (1.53 μ m) σ_{12}	$5.8 \times 10^{-21} \text{ cm}^2$
Signal emission cross	
section (1.53 μ m) σ_{21}	$6.1 \times 10^{-21} \text{ cm}^2$
ESA cross	
section (1.48 μ m) σ_{24}	$0.85 \times 10^{-21} \text{ cm}^2$
Cooperative UC C_2	$(2.65 \times N_{Er}/10^{20} + 3.38)$
	$\times 10^{-18} \text{ cm}^{3}/\text{s}$
Second-order cooperative	
UC C ₃₇	$(1 \times 10^{-17} - 1 \times 10^{-15}) \text{ cm}^3/\text{s}$
${}^{4}I_{13/2}$ state lifetime τ_2	7.8 ms
${}^{4}I_{11/2}$ state lifetime τ_{3}	30 µs
$^{4}I_{9/2}$ state lifetime τ_{4}	1 ns
Refractive index of erbium	
doped active regions	1.64
Refractive index of silicon waveguide	3.48

- The normalized pump and signal intensity profiles, i.e., Ψ_s(x, y) and Ψ_p(x, y), for the fundamental transverseelectric (TE) mode are calculated in erbium-doped Al₂O₃ using the finite element method by COMSOL Multiphysics, maintaining the intensity distributions for both signal and ASE. Here, TE polarization is defined as the one in which the electric field vector is parallel to the waveguide slab.
- 2) At each step in the propagation direction, by solving (1)–(4), Er^{3+} populations at energy levels N_1-N_4 at each point in the erbium region are calculated.
- 3) Absorption and emission coefficient $\gamma_{12}(z, v_s)$, $\gamma_{21}(z, v_s)$, and $\gamma_p(z)$ are calculated using (12)–(14), through $\Psi_s(x, y)$, $\Psi_p(x, y)$, N_1 and N_2 .
- Numerical integration of propagation equations (8)–(9) is achieved by the Runge–Kutta based iterative procedure to obtain P_p(z) and P_s(z) at each step.
- 5) The total pump power is then further attenuated by the nonlinear pump power loss in silicon (ΔP_{si}) at each propagation step as calculated using (19).

III. DEVICE GEOMETRY

The device structures of silicon rib waveguides with multiple erbium-doped Al₂O₃ regions used are depicted in Fig. 3. In these structures, the silicon (refractive index $n_{si} = 3.48$) rib waveguide acts as a platform to hold rectangular lowindex erbium-doped Al₂O₃ ($n_{Er-Al_2O_3} = 1.64$) regions. In particular, two groups of waveguide geometries (WG1 and WG2 of Fig. 3) are investigated in detail to evaluate the effects of power distribution on nonlinear losses. The waveguide height was fixed at 0.8μ m and waveguide widths of 0.6 μ m and 1 μ m were selected for the groups WG1 and WG2, respectively. Selection of waveguide width was in accordance with the desired single-mode condition in both polarizations, as well as accommodating several low-index erbium-doped active regions [23]. Minimal trench widths and spacings between the trenches were designed as 0.1 μ m owing to practical fabrication precision, complexity, and cost. As a result, the maximum number of trenches was constrained for different waveguide width designs. The length of the waveguide was selected to be 1 cm due to limitations of practical fabrication precision, complexity, and cost. For WG1 in Fig. 3(a), two subgroups, i.e., single trench (WG1A) and two trenches (WG1B), configurations were proposed and analyzed. For WG2 in Fig. 3(b), two trenches (WG2A) and four trenches (WG2B) were proposed and analyzed. In the above-mentioned designs, a maximum of two trenches (WG1B) for WG1 were designed with a total active area of 0.2 μ m \times 0.7 μ m and a maximum of four trenches (WG2B) for WG2 were designed with a total active area of 0.4 μ m \times 0.7 μ m. For the same silicon waveguide dimensions (width and height), the total erbium-doped active area remained the same for different numbers of active regions (subgroup A or B), showing the effects of trench numbers. By using waveguides with different widths (WG1 and WG2), our goal was to demonstrate how the power distributions and loss mechanisms in the waveguide are affected.

IV. RESULTS AND DISCUSSION

Signal amplification in the proposed structures strongly depends on erbium concentration as well as nonlinear loss mechanisms such as cooperative UC, ESA in erbium-doped Al₂O₃ regions, and free-carrier absorption in silicon, as illustrated in Fig. 1. Among these loss mechanisms, linear loss of the waveguide is the major reason for amplifier performance degradation and mainly results from the losses at the lateral walls of the waveguide during the etching process. However, since linear loss is not a design parameter, but rather a consequence of the fabrication processes, it will not be discussed in these calculations for device characterization and optimization.

A. Effects of Power Confinement and Population Inversion

As net gain and nonlinear losses are highly influenced by the optical intensity distribution, the impact of intensity distribution facilitated by different geometries (number of trenches) on each loss mechanism needs to be studied to determine the optical gain. The mode distributions in trenched waveguides are calculated using finite element method provided by COMSOL Multiphysics. From these simulations, the confinement factor and the distribution of normalized power intensity in erbium-doped Al₂O₃ regions are extracted for the fundamental TE mode, as illustrated in Fig. 4(a) and (b), respectively. In this structure, TE polarization is characterized as the one in which the electric field vector is perpendicular to the trenches to provide high power confinement in the Er^{3+} regions. To have a fair comparison, the total volume of the Er^{3+} -doped active regions is set to be equal for the same waveguide group (same waveguide width).

In the analyzed devices, the degree of Er^{3+} population inversion in the slot regions is plotted along the length of



Fig. 3. (a) Illustrations of WG1 of waveguide of dimensions 0.6 μ m width, 0.8 μ m height, 0.1 μ m trench width, WG1A (single trench), and WG1B (two trenches); waveguide length = 1 cm. (b) Illustrations of WG2 of waveguide of dimensions 1 μ m width, 0.8 μ m height, 0.1 μ m trench width, WG2A (two trenches), and WG2B (four trenches); waveguide length = 1 cm.

the waveguide as shown in Fig. 5. The population inversion of Er^{3+} is integrated over the thickness of the waveguide. The dip in the center of the waveguide is due to ESA. The effect of ESA does not decrease over the length of the waveguide since linear loss in the waveguide is not considered. If linear waveguide loss is considered, the effect of ESA decreases along the length of the waveguide as a result of absorption by Er^{3+} and waveguide losses, resulting in little or no dip at the end of the waveguide [24].

B. Effects of Second-Order UC and Er^{3+} Concentration

Second-order UC is known to cause fluorescence at multiple wavelengths, such as 520 nm green emission due to transitions from the ${}^{2}H_{11/2}$ state to the ground state, along with fluorescence at 545, 660, and 850 nm [16], [17], [25], [26]. In particular, second-order UC is the leading detrimental effect at high erbium concentrations of $\geq 1 \times 10^{21}$ cm⁻³ [16]. A theoretical estimate of the effects of the second-order UC has been performed by varying the second-order UC coefficient C_{37} based on the reported value [17] in (1)–(4). As the intensity distribution in the waveguide is nonuniform and the C_{37} is intensity-dependent, population inversions are calculated at different geometrical locations [points A–I in

WG1 group	WG1A	WG1B
Power confinement in erbium doped Al_2O_3	16.04%	27.52%
WG2 group	WG2A	WG2B
Power confinement in erbium doped Al_2O_3	22.82%	33.85%



Fig. 4. (a) Power confinement factors of different proposed geometries. (b) Illustration of normalized power intensity profile (per m^2) of a twotrenched waveguide (WG1B) in WG1 group.



Fig. 5. Degree of inversion along the length of the waveguide. Calculations are made for WG1B, pumped at 100 mW at 1.48 μ m (waveguide length = 1 cm).

Fig. 6(a)] in the erbium-doped Al₂O₃ region. Here, each point corresponds to different pump intensities, as indicated by the contour lines in Fig. 6(a). By using $C_{37} = 8 \times 10^{-16}$ cm³/s [17], we have shown that the population inversions decrease by at least 30% as depicted in Fig. 6(b). Furthermore, second-order UC may not lead to an overall net population inversion depending on the intensity distributions in the geometry, as indicated in Fig. 6(b) and in [16], [17], and [26].

Although high erbium solubility in Al₂O₃ up to 3×10^{21} Er/cm³ is achievable [26], steady-state erbium population contributing to signal amplifications is greatly reduced by the second-order UC. Thus, high erbium concentration (e.g., $\ge 1 \times 10^{21}$ Er/cm³) is not suitable for this [26]. High-order emission peaks are strongly reduced at low erbium concentration of



Fig. 6. (a) Contour plot of pump intensity distribution within erbium-doped Al₂O₃. Intensity distributions at exemplary points A-I show the estimated population inversion at locations with different pump intensities. Numbers on the contour lines represent intensity (kw/cm⁻³) (e.g., 2000 represents 2000 kw/cm⁻³). (b) Population inversion calculated at points A-I with respect to different pump intensities and different values of C₃7.

 $\leq 5 \times 10^{20}$ Er/cm³, thus second-order UC can be neglected if the device is strategically fabricated and engineered [16], [17]. Subsequently, ESA and UC (excluding second-order UC in this paper) are the detrimental effects that need to be incorporated at lower concentrations. Optimum erbium concentration of 5×10^{20} Er/cm³ is used in the following analysis to provide high optical gain as depicted in Fig. 7, without second-order UC degradation.

C. ESA and Cooperative UC

Although there have not been direct measurements of ESA in erbium-doped Al₂O₃ with 1.48 μ m pump wavelength, the ESA process from ⁴I_{13/2} to ⁴I_{9/2} is taken into account based on the experimental results for the gain calculation of erbium-doped planar waveguides and erbium-doped fiber amplifiers [11], [16], [17], [26], [27]. The ESA process is observed at lower erbium concentrations and explained as a phonon-assisted process [16]. Thus, the ESA cross section determined from the experimental data is host-material dependent and



Fig. 7. Signal gains of WG1A versus pump powers of different erbium doping concentrations ($5 \times 1020 \text{ cm}^{-3}$, $2 \times 1020 \text{ cm}^{-3}$, $1 \times 1020 \text{ cm}^{-3}$, $5 \times 1019 \text{ cm}^{-3}$), waveguide length = 1 cm.



Fig. 8. Signal gains of WG1A with no ESA and UC, no UC, no ESA, and with both ESA and UC (waveguide length = 1 cm).

nearly a factor of 10 smaller than the absorption cross section for excitation of the first excited state [16].

In this paper, by using the physical parameters determined experimentally [11], [16], ESA is found to deplete the optical gain at high pump intensities, especially at 1.48 μ m pump wavelength. As illustrated in Fig. 8, ESA is the strongest effect that reduces the gain at pump powers above 20 mW. Similarly, the cooperative UC process is affected by pump intensity at higher erbium concentration, and gain saturation is introduced at low pump power values, as shown in Fig. 8. The gain profile generated by single-trench (WG1A) and double-trench (WG1B) waveguides are shown in Fig. 9(a). These results indicate that device performance will be limited to low pump operations due to nonlinear losses in erbium-doped regions. Up to 0.38 dB/cm more signal gain is provided in the double-trench design, as shown in Fig. 9(a).

Since linear loss is not considered in this analysis, nonlinear losses in the structures are calculated as

$$L_{NF} = G_{sat} - G \tag{20}$$



Fig. 9. (a) Signal gain and nonlinear losses versus pump power for WG1A and WG1B (waveguide length = 1 cm). (b) Signal gain and nonlinear losses versus. pump power for WG2A and WG2B (waveguide length = 1 cm).

where G_{sat} is the saturation gain without linear and nonlinear losses and G is net optical gain considering nonlinear losses. G_{sat} is estimated as

$$G_{sat} = 4.43 \times (\sigma_{21}N_2 - \sigma_{12}N_1)\Gamma L$$
 (21)

where σ_{12} and σ_{21} are the Er³⁺absorption and emission cross sections, Γ is the power confinement factor, and L is the waveguide length. As shown in Fig. 9(a), nonlinear losses are increased in double-trench structures. It is mainly caused by the nonuniformity of the power intensity profile in multitrench structures. The improved signal gain of the double-trench waveguide is mainly due to the improved power confinement factor, which improves the saturated gain without linear and nonlinear losses. Larger waveguides can accommodate more erbium-doped regions and have higher power confinement, as seen in Fig. 6(a). The maximum achievable signal gain is improved in the larger waveguides by at least 0.3 dB/cm, comparing the four-trench configuration (WG2B) in Fig. 9(b) with the two-trench configuration (WG1B) in Fig. 9(a). With no linear loss in the system, as shown in Fig. 9(b), the four-trench design can achieve the maximum gain of 1.1 dB/cm.



Fig. 10. Noise figure versus input pump power for WG1B and WG1A (waveguide length = 1 cm).

Linear loss due to scattering losses on the slot sidewalls is another parameter that should be considered. However, since linear loss is strictly a fabrication-dependent parameter, it is intentionally left out in this analysis. In the presence of linear loss, the effect of ESA decreases along the length of the waveguide as a result of absorption by Er^{3+} and waveguide losses. In addition, net gain may deplete or vanish depending on the magnitude of the linear loss.

D. Noise Figure

The noise figure of the analyzed devices can be easily obtained once the ASE has been spectrally resolved. By considering only signal–spontaneous emission beat noise, the noise figure can be expressed as [28]

$$NF(dB) = 10\log_{10}\left[\frac{1}{G(z)} + \frac{P_{ASE+}(z, v_s)}{G(z)hv_s\delta v}\right].$$
 (22)

G(z) is the signal gain, v_s is the signal frequency, and δv is the width of the frequency slot used to compute $P_{ASE+}(z, v_s)$. In this paper, the range of ASE spectrum calculated is 1.45– 1.65 μ m and $\delta v = 128$ GHz. The noise figure analysis only applies to the unsaturated gain regime and evaluates the performance of the integrated amplifiers. Due to the nonuniformity of the pump profile, the signal-to-noise ratio is degraded in the multitrench waveguide as shown in Fig. 10. The multitrench waveguide has higher power confinement and population inversion, which leads to more spontaneous decay.

E. Free-Carrier Absorption

At high pump powers, optical loss induced by free-carrier absorption in silicon cannot be neglected [21]. As the maximum pump power in the proposed waveguide structures are limited by the ESA and UC, we expect to have limited free-carrier losses in the silicon region. For pump power lower than 100 mW (to minimize ESA and UC effects), the signal gain tends to decrease to 3×10^{-4} dB/cm, as shown in Fig. 11. These results indicate that the effect of the free-carrier loss is negligible for pump powers less than 100 mW.

In addition, photon absorption in silicon due to the emissions from N_3 and N_4 states in the erbium-doped Al_2O_3



Fig. 11. Signal gain versus pump power for WG1A with and without freecarrier absorption in silicon (waveguide length = 1 cm).

regions is also considered. The upper state emissions disappear immediately in the silicon nearby and may become the free carriers, because the absorption coefficient of the Si layer is large, about 1000 cm⁻¹. Given the erbium ion population in N₃ and N₄ states (approximately 10^{18} cm⁻³), lifetime of N₃ and N₄ states (approximately in the order of hundreds of microseconds) and the electron–hole recombination time (approximately in the order of nanoseconds), the generated free-carrier density is estimated to be approximately 10^{10} /cm³ and thus it is negligible.

V. CONCLUSION

In conclusion, the effects of design geometries and nonlinear losses on gain in silicon waveguides with erbium-doped Al_2O_3 regions were studied. The performance of silicon waveguides with erbium-doped Al_2O_3 regions was limited by the UC and ESA. Second-order UC was found to be one of the detrimental effects that prevent signal gain at high erbium concentrations. Distributing erbium ions over multitrench areas improves the gain-to-nonlinear-loss ratio. Thus, 0.38 dB/cm higher signal amplification was generated when compared to a single-trench configuration. Linear losses of the analyzed devices are mainly due to the fabrication processes. However, linear losses cannot be ignored in final designs, and thus the fabrication processes need to be improved to provide ultimate high signal amplification.

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